ELECTROMAGNETIC SENSOR ARRAY FOR TRAIN WHEEL DETECTION

Patricio G. Donato, Jesús Ureña, Manuel Mazo, Juan Jesús García, Fernando Álvarez*

Department of Electronics, University of Alcalá, Alcalá de Henares, 28871 (Spain)
{donatopg, urena, mazo, jesus}@depeca.uah.es
*Dept. of Electronics and Electromechanical Eng., University of Extremadura, Cáceres, 10071 (Spain)
fafranco@unex.es

ABSTRACT

This work proposes a train wheel detection system based on an electromagnetic sensor array. This sensor array is distributed in points along a section of the railway. Each sensor of the array is located next to the rail without additional electronics in this place. All the electronic signal process equipment is concentrated at a place far away the rail. To compensate the attenuation in the wires and the external noise, it is proposed to use a signal coded with Golay sequences. Combining the properties of the Golay sequences with the characteristics of the proposed sensor array, it is possible to detect the passage of the train by scattered points, and also to recognize the sense of circulation and estimate the speed of the train.

2. DESCRIPTION OF THE COMPLETE SYSTEM

The block diagram of the detector system is shown in fig.1. The emitted signal is coded with a pair of Golay complementary sequences of length N bits. Subsequently the signal is digitally modulated and amplified, and then is used to excite the array sensor. The sensor coils of the array are interconnected in series among them and with the circuit by means of an AC bridge. The sensor coils are placed next to the rail, in determined detection points, and are mounted in a fixed position and orientation. The study about the magnetic field and the influence of the wheel of train it was developed in [4]. In the steady state (when the rail is free) the bridge is balanced and the voltage between the midpoints of the bridge is null (or nearly null). The bridge turns unbalanced when a train wheel is above any of the sensor coils. In this case a signal appears that is treated by the process unit. This signal has the same waveform of the signal used to excite the bridge, but is attenuated and with added noise. The signal is analogically conditioned and digitalized. Then, the signal is digitally demodulated and at the end is correlated. The output of the correlation will give the characteristics peaks of the Golay sequences, which will be validated by the peak detector. When a train passes by a detection point, the system will identify the signal by means of the peaks, even in presence of noise and interferences.

1. INTRODUCTION

Almost all commercial electronic train wheel detectors are based on RLC circuits placed next to the rail and tuned in a fixed frequency. When a train passes above the detector, the impedance changes due to the variations in the magnetic permeability and flux density produced by the wheel of the train. This causes variations in the current across the RLC circuit, which are easily measurable. Other systems propose a fail safe wheel train detector using two coils working as emitter and receiver respectively [1] [2]. The detector senses the wheels as changes in the magnetic coupling between the coils placed at either rail sides. The system described in [3] also uses a pair of coils as emitter and receiver, but it works processing a Barker code in a continuous way. The signal process is made by means of the application of the correlation function to the received signal. This process takes advantage of the Barker code correlation properties, which allows distinguishing the Barker code in spite of the external noise or the attenuation of the channel. The problem of Barker codes is their maximum length of 13 bits that limits the desired SNR of work.

The proposed system in this work uses pairs of emitter and receiver coils as track sensors. It is based on [4], but instead of two, it proposes to use an array of n sensor coils located in n detection points. With this, it is possible to detect the passage of the train by scattered points of a railway section. Moreover, we suggest detecting the direction of the circulation and estimate the speed using the same sensor array. To compensate the external noise, it uses a signal coded with Golay complementary sequences [7], and taking advantage of this, the electronic process unit is moved away of the railway. The correlation will be applied for the process of the signal. It can be implemented in reprogrammable logic systems [5], using efficient Golay correlators [6].

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3. SENSOR CIRCUIT

The sensor coils are distributed in the railway in specific detection points (fig. 2). Besides that, the sensors coils are interconnected in series with their homologous terminals connected in a different way, forming a two-wire sensor (fig. 3). The sensor circuit is based on an AC bridge, controlled by the coded signal with the sensor coils in one of the branches. The particular AC bridge used is an implementation of the Maxwell bridge (fig. 4). When the system is in steady state (the rail is free), the bridge is balanced and the voltage between the midpoints of the two branches is null. When the wheel of the train passes by the place where there are the coils, the coefficient of coupling falls in the same ones. This causes an unbalance in the bridge, and a differential voltage appears between the two branches, with the same waveform that the coded signal from the power circuit. In order to detect the direction of circulation, at least two points of detection are needed. We use \( n \) detection points, not only to know the direction, but also to measure the speed and the position in a certain section of the railway. The \( n \) pairs of emitter & receiver coils are connected in series (fig. 3), making an array of inductances that has a total inductance \( L_x \). The changes in \( L_x \) will be different according to which pair of coils is affected. The mathematical expression of \( L_x \) is:

\[
L_x = \sum_{i=1}^{n} (L_{Ti} + L_{Ri} + 2M_i) \tag{1}
\]

\( M_i \) is the mutual inductance of each pair of coils, and its sign depends on the connection of the homologous points of them:

\[
M_i = \pm k_i \cdot \sqrt{L_{Ti} \cdot L_{Ri}} \Rightarrow k_i \in [0;1] \tag{2}
\]

The coefficient of coupling \( k_i \) between the receiver and the emitter coil is the ratio of the flux lines produced by the emitter coil, that link the receiver coil to the total flux produced by the emitter coil. Therefore, the variations of \( k_i \) will be proportional to the variations in the magnetic flux. Supposing the inductances \( L_{Ti} \) and \( L_{Ri} \) are not affected for the wheel, the attenuation ratio of \( M_A \) is:

\[
r_{\phi_A} = \frac{\Phi_{\text{free}}}{\Phi_{\text{busy}}} \Rightarrow r_{\phi_A} \in [0;1] \tag{3}
\]

The passage of a wheel by the detection point \( A \) affects only the mutual inductance of the pair of coils corresponding to that point and not over the coils placed at other detection points. So the mathematical expressions of the \( L_x \) in the cases of the free rail and the busy rail are:
\[ L_{X}^{\text{free}} = \sum_{i=1}^{n} (L_{Ti} + L_{RI} + 2M_i) \] (5)

\[ L_{X}^{\text{bus}} = \sum_{i=1}^{n} (L_{Ti} + L_{RI} + 2M_i) + 2M_{A}^{\text{bus}} \] (6)

\[ L_{X}^{\text{free}} = \sum_{i=1}^{n} (L_{Ti} + L_{RI} + 2M_i) - 2M_{A}^{\text{free}} + 2M_{A}^{\text{bus}} \] (7)

The values of \( M_A \) for the two cases are related by means of the expression of \( r_{MA} \) (4). Then, the attenuation of the equivalent inductance \( L_x \) is:

\[ \Delta L_{X} = L_{X}^{\text{free}} - L_{X}^{\text{bus}} = 2(M_{A}^{\text{bus}} - M_{A}^{\text{free}}) \] (8)

\[ \Delta L_{X} = 2(r_{MA} - 1) \cdot M_{A}^{\text{free}} \] (9)

Therefore the variations in the equivalent inductance are directly proportional to the variations in the magnetic flux. Due to the magnitude of the attenuation ratio of the mutual inductance \( M_A \) is comprised between 0 and 1 (3) (4), the range of variation of \( \Delta L_{X} \) is:

\[ \Delta L_{X} \in [-2M_A;0] \] (10)

The value of \( M_A \) can be bigger or lower than zero, this depends on the connection of the homologous points of the coils at the detection point A. It remains only to estimate the module and phase of the signal between the midpoints of the AC bridge. From the bridge of fig. 4, the expression of \( V_{BA} \) can be calculated as follows:

\[ V_{BA} = \frac{Z_{1}Z_{3} - Z_{2}Z_{X}}{Z_1 + Z_2} \Rightarrow Z_X = f(L_X) \] (11)

\[ V_{BA} = \frac{\left[ R_{1}R_{2} + R_{2}R_{3} + j\alpha R_{L}(R_{1} + R_{2})\right]}{\left[ R_{1} + R_{2} + R_{1} + R_{3}\right]} - \frac{1}{\alpha} \frac{\alpha R_{L}(R_{1} + R_{2})}{\left[ R_{1} + R_{2} + R_{1} + R_{3}\right]} + j\alpha R_{L}(R_{1} + R_{2}) + CR_{L}(R_{1} + R_{2}) \] (12)

With (12), it is possible to know the variations in the output signal related to the inductance, and therefore, related to the magnetic flux attenuation. To illustrate this, eight pairs of coils distributed in eight detection points are used (fig. 5) as example. In fig. 6 the module and phase of \( V_{BA} \) (12) are represented in a range of values of \( L_X \) comprised between 50mH and 70mH, and considering an input signal of amplitude ±1. The design values used to balance the bridge are:

- \( R_1 = R_2 = 66.7 \Omega \)
- \( R_3 = R_4 = 9 \Omega \)
- \( C_1 = 0.1 \mu F \)
- Frequency of modulation = 50 kHz
- Inductance in balance point = 60 mH

The values marked in fig. 6 are resumed in Table 1. All the data it was obtained simulating the system in Simulink, with the Power System Blockset [10]. They are referred to the changes in \( L_X \) when the train is above any of the detection points (DP’s, from 1 to 8). The values of \( \Delta L_X \) were selected to produce approximately proportional variations in the module of \( V_{BA} \). The values of \( k \) and \( r_{MA} \) are equal because all the pairs of coils are physically mounted in the same way, and the effect of the passage of the wheel is similar in all the detection points. \( M_{A}^{\text{free}} \) was designed based on (9) and the sum of all the values when the rail is free is null (5).

![Fig. 5: Distribution of eight pairs of coils in the railway.](image)

![Fig. 6: Module and phase of \( V_{BA} \) related to \( L_X \).](image)

<table>
<thead>
<tr>
<th>DP</th>
<th>( r_{MA} )</th>
<th>( k )</th>
<th>( M_{A}^{\text{free}} ) (mH)</th>
<th>( \Delta L_X ) (mH)</th>
<th>( L_{X} ) (mH)</th>
<th>( A V_{BA} )</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>0.1</td>
<td>0.55</td>
<td>+3.30</td>
<td>-6.0</td>
<td>54.0</td>
<td>0.026</td>
</tr>
<tr>
<td>2</td>
<td>0.1</td>
<td>0.55</td>
<td>+2.48</td>
<td>-4.5</td>
<td>55.5</td>
<td>0.020</td>
</tr>
<tr>
<td>3</td>
<td>0.1</td>
<td>0.55</td>
<td>+1.70</td>
<td>-3.0</td>
<td>57.0</td>
<td>0.013</td>
</tr>
<tr>
<td>4</td>
<td>0.1</td>
<td>0.55</td>
<td>+0.83</td>
<td>-1.5</td>
<td>58.5</td>
<td>0.007</td>
</tr>
<tr>
<td>5</td>
<td>0.1</td>
<td>0.55</td>
<td>-0.83</td>
<td>+1.5</td>
<td>61.5</td>
<td>0.007</td>
</tr>
<tr>
<td>6</td>
<td>0.1</td>
<td>0.55</td>
<td>-1.70</td>
<td>+3.0</td>
<td>63.0</td>
<td>0.013</td>
</tr>
<tr>
<td>7</td>
<td>0.1</td>
<td>0.55</td>
<td>-2.48</td>
<td>+4.5</td>
<td>64.5</td>
<td>0.018</td>
</tr>
<tr>
<td>8</td>
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<td>0.55</td>
<td>-3.30</td>
<td>+6.0</td>
<td>66.0</td>
<td>0.024</td>
</tr>
</tbody>
</table>
4. PROCESS & CODIFICATION OF THE SIGNAL

4.1. Codification of the signal with Golay sequences

Given a pair of Golay complementary sequences \(\{a, b\}\) of length \(N\) [7] [4] (fig. 7), the sum of the autocorrelations of both sequences is a delta of Dirac \(\delta(n)\) of amplitude \(2N\). This is even fulfilled if the sequences are superposed [8], as long as the separation between the superposed sequences is bigger than zero. That is to say, that at least there is an element of separation between superposed sequences. In this work, it used a sequence superposition of \(N/2\) bits:

\[
A(n) = \sum_k a(n-k \cdot N/2) \forall k
\]

\[
B(n) = \sum_k b(n-k \cdot N/2) \forall k
\]

\[
C_{ab} = 2N \sum_k \delta(n-k \cdot N/2) + \frac{C_{aa} + C_{bb}}{4}
\]

In the sum of correlations (14) it was assumed a Gaussian noise, whose contribution is \(2N\) times lower than the amplitude of the deltas. The superposed sequences \(A(n)\) and \(B(n)\) are periodic signals with a period of \(N/2\). Moreover, \(N/4\) of the values of \(A(n)\) and \(B(n)\) are null. The fact that \(N/4\) of the values are null allows us to send in the time \(A(n)\) and \(B(n)\) by the same channel, with a displacement of one of them \(N/4\) respect to the other.

In fig. 7 it can be seen a pair of Golay complementary sequences \(\{a, b\}\), of length \(N=32\) bits. In fig. 8, the same sequences \(\{a, b\}\) are present but superposed \(N/2\) bits in a continuous emission, as is mathematically expressed in (13). In this figure it can be seen how the continuous emission of both sequences produces a periodic signal, with a period of \(N/2\) bits (in this case, 16 bits), and \(N/4\) bits are zero.

![Fig. 7: Golay complementary sequences of length \(N=32\).](image1)

![Fig. 8: Golay sequences superposed \(N/2\) bits (with \(N=32\)).](image2)

4.2. Implementation of the system

With the generated superposed sequences, the signal sent to the emitter coil is modulated digitally by means of a BPSK modulator. So the signal adapts to the central frequency of work looked for the system, and it has a narrow bandwidth. When the wheel of the train is at a detection point (above a pair of coils), the flux in the receiver coil is approximately attenuated 10 times (\(k=0.1\)) [9]. Then a differential voltage appears in the bridge and is processed by the Golay correlator and the
peak detector. In fig. 5 it is shown the disposition of the coils in the railway (eight in this case). In fig. 10 can be seen the results of correlation using Golay sequences of 32 bits, without noise, and supposing a train circulating through all the detection points. In fig. 11 the results of correlation with the same conditions of process and codification are displayed, but with a SNR=−9dB. All the process and codification it was realized in Matlab [10]. In this figure the effect of the passage of the wheel above the different sensors can be clearly identified by means of the different amplitudes and signs of the correlation peaks.

![Fig. 10: Correlation peaks without external noise.](image1)

![Fig. 11: Correlation peaks with SNR=−9dB.](image2)

### 5. CONCLUSIONS

In this work we have presented an electronic wheel detector with a sensor array capable to detect the position, sense of circulation and speed of the train at different points. This sensor array is composed of pairs of coils, which are interconnected using only two wires and are part of an AC bridge, which contribute to improve the sensibility. The detector uses a continuous emission of a coded signal to work with a low SNR. With this technique, the signal attenuation in the wires is compensated and immunity to the interferences is obtained. The signals can be detected with noise at different or the same frequency that one used in the modulation. This allows us to place the sensors in the rail without the signal process equipment in the same place. It makes easier the maintenance tasks and improves the robustness of the external equipment. The implementation of Golay sequences on FPGA, permits to work in real time and with reliability.

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### 7. REFERENCES


